

5 DECISION FEEDBACK AND WHITENING RECEIVERS

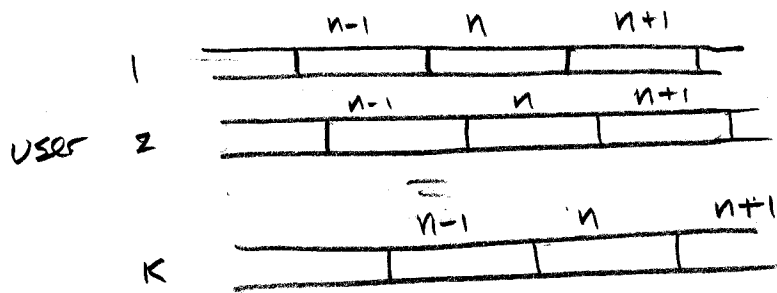
- To this point, the user data values b could just as well have been analog. We haven't exploited the fact that they are discrete and finite to reduce interference.
- The analogy between linear MUD and linear equalizers makes the next step irresistible — decision feedback.
- Recall from single-user equalizers that DF removes about half of the ISI — if the feedback decisions are correct (DA, data aided). If decision directed (DD), then some degradation through error propagation.
- If we could remove half of the MAI, we might have more degrees of freedom left for diversity combining.
- Several variants of DF, with significantly differing performance.

5.1 Unmodified Decision Feedback

- After MF, Rake/MRC, we have

$$\underline{y} = \underbrace{C^T R C A}_{H} \underline{b} + \underline{\eta} = H \underline{b} + \underline{\eta}, \quad R_{\eta} = N_0 C^T R C$$

The $NK \times NK$ matrix H is block tridiagonal, reflecting the overlapping pulses



- Why not make decisions in order?

$$b_1(n-2), b_2(n-2), \dots, b_K(n-2), b_1(n-1), b_2(n-1), \dots, b_K(n-1), b_1(n), b_2(n), \dots$$

- For any $b_k(n)$ we have decision variable

$$\begin{aligned} \tilde{y}_k(n) = & h_{1,0}^{(k,n)} b_{1,0} + \dots + h_{k-1}^{(k,n)} b_{k-1} + h_k^{(k,n)} b_k(n) \\ & + h_{k+1}^{(k,n)} b_{k+1} + \dots + \eta_k(n) \end{aligned}$$

where $h_k^{(k,n)}$ is element of row $k+nK$ of H corresponding to $b_k(n)$.

- Subtract earlier decisions $\tilde{b}_k(n')$:

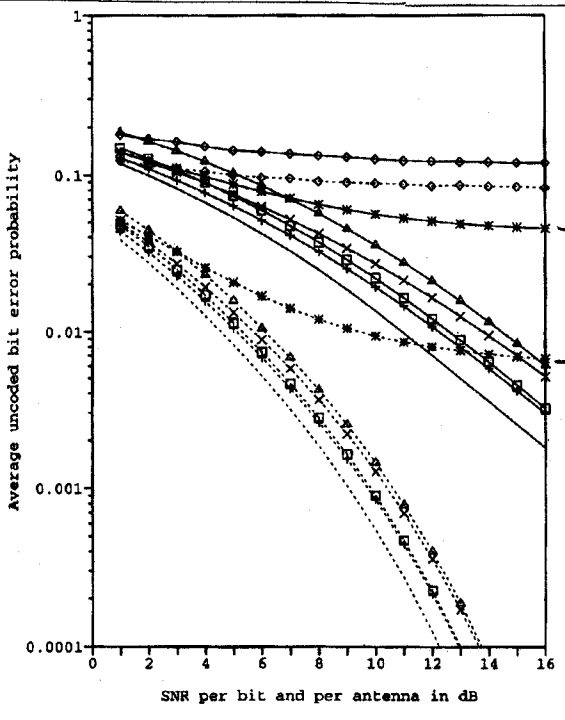
$$\begin{aligned} \tilde{y}_k'(n) = & h_{1,0}^{(k,n)} e_{1,0} + \dots + h_{k-1}^{(k,n)} e_{k-1}(n) \\ & + h_{k,n}^{(k,n)} b_k(n) + \eta_k(n) \\ & + h_{k+1}^{(k,n)} b_{k+1}(n) + \dots \end{aligned}$$

where $e_k(n') = b_k(n') - \tilde{b}_k(n')$

• But it turns out not to work very well.
 Results from [Jung95], repeated from Section 4.4
 K = 8 users $N_c = 14$ chips $N = 20$ sym/burst

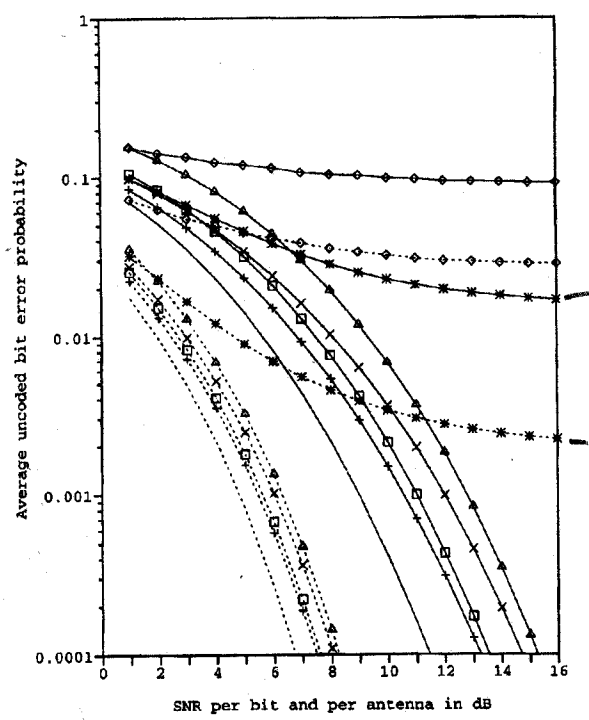
L = 2

L = 27



DMF of (36), $K_a = 1$	—
MMSE-BDFE, $K_a = 1$	+
ZF-BDFE, $K_a = 1$	□
MMSE-BLE, $K_a = 1$	×
ZF-BLE, $K_a = 1$	*
DMF-BDFE, $K_a = 1$	•
DMF of (35), $K_a = 1$	◊
DMF of (36), $K_a = 1$	◊
MMSE-BDFE, $K_a = 1$	+
ZF-BDFE, $K_a = 1$	□
MMSE-BLE, $K_a = 1$	×
ZF-BLE, $K_a = 1$	*
DMF-BDFE, $K_a = 1$	•
DMF of (35), $K_a = 1$	◊

We are doing DMF-BDFE



It's a lot better than simple Rake ($\tilde{b}_k(n) = \text{sgn}[\tilde{r}_k(n)]$), but there is still a significant floor.

• What is the problem?

— We still have MAI from bits yet undetected.

— The noise samples in η are correlated, so $\tilde{r}_k(n')$ samples can help us in $b_k(n)$ decision through noise minimization as well as MAI cancellation — an unrealized possibility, since we use only $\tilde{r}_k(n)$.

• Note the requirement to know H (channel and signal correlations) in order to subtract interference!