

2.2 Second Order Statistics of Complex Signals

2.2.1

- In this section, we look at 2nd order stats — power, autocorrelation, power spectrum — to see how real bandpass and complex envelope stats are related.
- First, energy. The energy of $\tilde{v}(t)$ is half that of its complex envelope $v(t)$:

$$\int \tilde{v}^2(t) dt = \frac{1}{2} \int |v(t)|^2 dt$$

or by Parseval

$$\int |\tilde{v}(f)|^2 df = \frac{1}{2} \int |v(f)|^2 df$$

Why?

Because the former is the actual transmitted energy, it is common in communications to write

$$E_v = \frac{1}{2} \int |v(t)|^2 dt = \frac{1}{2} \int |v(f)|^2 df$$

- Next, random processes. Define correlation functions

$$R_x(\tau) = \frac{1}{2} E[x(t) x^*(t-\tau)]$$

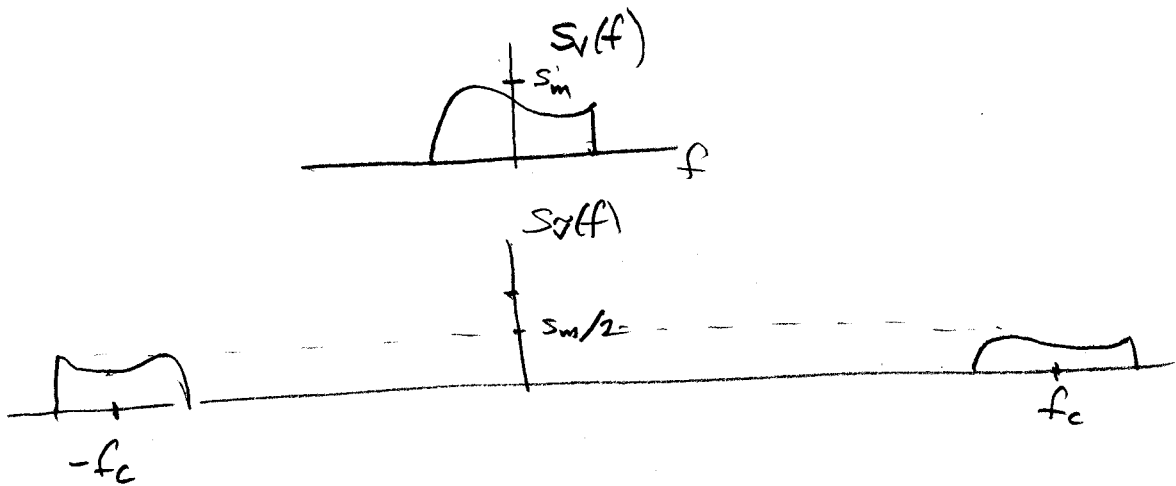
$$R_{xy}(\tau) = \frac{1}{2} E[x(t) y^*(t-\tau)]$$

$$P_x = R_x(0)$$

- Statistics of the corresponding bp process are given by modulation relationships like those linking $v(t)$ and $\tilde{v}(t)$:

$$R_{\tilde{v}}(\tau) = E[\tilde{v}(t)\tilde{v}(t-\tau)] = \text{Re} [R_v(\tau) e^{j2\pi f_c \tau}]$$

$$S_{\tilde{v}}(f) = \mathcal{F}[R_{\tilde{v}}(\tau)] = \frac{1}{2} S_v(f) + \frac{1}{2} S_v(-(f+f_c))$$



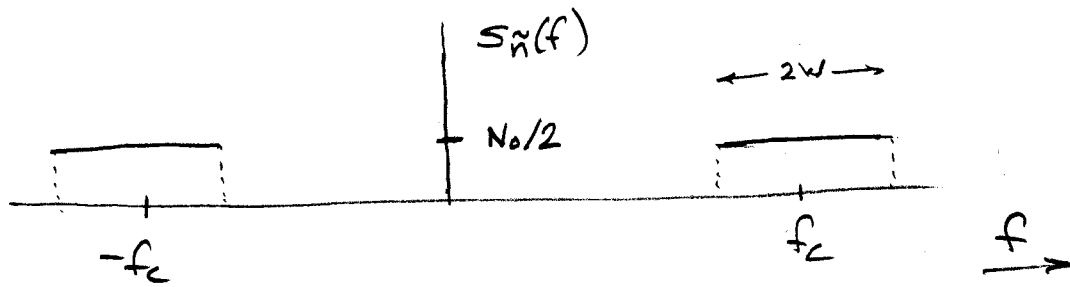
- $S_v(f)$ is real Why?

but not necessarily symmetric. Why not?

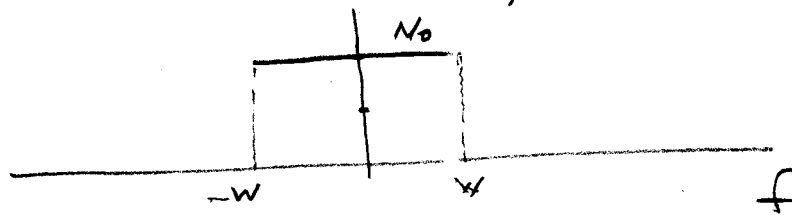
Symmetric if and only if $v_r(t)$ and $v_i(t)$ are uncorrelated. Why?

- White noise, in particular...

The real bandpass signal $\tilde{u}(t)$ has PSD



and complex envelope PSD



Are the powers the same?

- More detail Appendix B2

- Projection of white noise onto set of functions $u_1(t), \dots, u_N(t)$, not necessarily orthonormal.

$n(t)$ complex envelope, $R_n(\tau) = N_0 \delta(\tau)$

$$n_i = \int n(t) u_i^*(t) dt \quad \text{zero mean noise}$$

$$\begin{aligned} \text{so } \sigma_i^2 &= \frac{1}{2} E[|n_i|^2] = \frac{1}{2} E \iint n(t) n^*(s) u_i^*(t) u_i(s) dt ds \\ &= N_0 \iint \delta(t-s) u_i^*(t) u_i(s) dt ds \\ &= N_0 \int |u_i(t)|^2 dt \end{aligned}$$

and covariance

$$\begin{aligned} \sigma_{ik}^2 &= \frac{1}{2} E[n_i n_k^*] = \frac{1}{2} E \iint n(t) n^*(s) u_i^*(t) u_k^*(s) dt ds \\ &= N_0 \int u_i(t) u_k^*(t) dt \end{aligned}$$

- In particular, noise components wrt orthogonal functions are uncorrelated.

- As a vector $\underline{n} = \begin{bmatrix} n_1 \\ n_2 \\ \vdots \\ n_N \end{bmatrix} = \int n(t) \underline{u}^*(t) dt = \int n(t) \begin{bmatrix} u_1(t) \\ \vdots \\ u_N(t) \end{bmatrix}^* dt$

Covariance matrix

$$\begin{aligned} R_n &= \frac{1}{2} E[\underline{n} \underline{n}^T] = \frac{1}{2} \iint \overline{n(t) n^*(s)} \underline{u}^*(t) \underline{u}^T(s) dt ds \\ &= N_0 \underbrace{\int \underline{u}^*(t) \underline{u}^T(t) dt}_{\Phi_u} = N_0 \Phi_u \end{aligned}$$

• Projection of white noise onto subspace spanned by $u_1(t), \dots, u_N(t)$:

$$\hat{n}(t) = \sum_{i=1}^N v_i u_i(t) = \underline{v}^T \underline{u}(t)$$

where coefficients $\{v_i\}$ are selected to minimise

$$\underline{\epsilon}_\perp = \int |n(t) - \hat{n}(t)|^2 dt$$

Solution

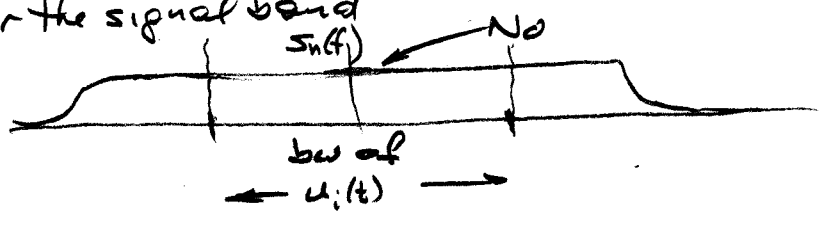
$$\underline{v} = \Phi_\perp^{-1} \underline{n}$$

- The proof is worth following through, since it illustrates classic linear approximation, but in complex formulation.

- Expand energy in the error *

$$\begin{aligned} \underline{\epsilon}_\perp &= \int |n(t) - \sum_{i=1}^N v_i u_i(t)|^2 dt = \int |n(t) - \underline{v}^T \underline{u}(t)|^2 dt \\ &= \underline{\epsilon}_n - \underline{v}^T \int n(t) \underline{u}^*(t) dt - \int \underline{u}^T(t) n^*(t) dt \cdot \underline{v} \\ &\quad + \underline{v}^T \int \underline{u}^*(t) \underline{u}^T(t) dt \cdot \underline{v} \\ &= \underline{\epsilon}_n - \underline{v}^T \underline{n} - \underline{n}^T \underline{v} + \underline{v}^T \Phi_\perp \underline{v} \end{aligned}$$

* But white noise has infinite power, so $\underline{\epsilon}_\perp, \underline{\epsilon}_n$ are infinite?!?
 Solution - dodge the issue and make $S_n(f)$ flat only over the signal band



— Minimise $\underline{\epsilon}_\perp$ wrt \underline{v} ; it's a scalar, so set the gradient $\nabla_{\underline{v}} \underline{\epsilon}_\perp = \underline{0}$. Next problem: coefficients \underline{v} are complex, but $\underline{\epsilon}_\perp$ is not analytic.

Solution 1: take grad wrt re and im parts separately

Solution 2: use the compact formalism of Appendix K

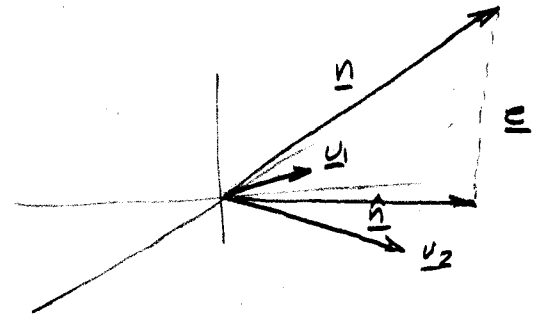
$$\begin{aligned} \nabla_{\underline{v}} \underline{\epsilon}_\perp &= \nabla_{\underline{v}} \underline{\epsilon}_n - \nabla_{\underline{v}} \underline{v}^\dagger \underline{\eta} - \nabla_{\underline{v}} \underline{\eta}^\dagger \underline{v} + \nabla_{\underline{v}} \underline{v}^\dagger \Phi_v \underline{v} \\ &= \underline{0} - \underline{0} - \underline{\eta} + \Phi_v \underline{v} = \underline{0} \end{aligned}$$

so $\boxed{\Phi_v \underline{v} = \underline{\eta}, \quad \underline{v} = \Phi_v^{-1} \underline{\eta}}$

These are the "normal equations": matrix of inner prods of basis functions (Gram matrix Φ_v) and vector of inner prods of function being approx'd and basis f's ($\underline{\eta}$).

— Important property: error $e(t) = n(t) - \hat{n}(t)$ is orthogonal to any vector in the subspace, including $\hat{n}(t)$. Easy proof: get vector of inner prods of $e(t)$ and basis functions.

$$\begin{aligned} \underline{e} &= \int e(t) \underline{v}^*(t) dt = \int n(t) \underline{v}^*(t) dt - \int \underline{v}^*(t) \underline{v}^\dagger(t) \underline{v} dt \\ &= \underline{\eta} - \Phi_v \underline{v} = \underline{0} \end{aligned}$$



$\underline{e} \perp$ all vectors in the subspace.

- Summary:

→ The coefficients \underline{v} and inner products \underline{n} are equivalent, since $\Phi, \underline{v} = \underline{n}$

→ Either is sufficient to characterise the projection $\hat{n}(t)$ of $n(t)$ onto $\{u_i\}$ space

→ That part of noise falling outside U space is orthog to everything in U

→ Components of $n(t)$ on basis vectors orthog to U are uncorrelated with \underline{n}

• Finally, if vector is complex Gaussian, then

$$p_{\underline{n}}(\underline{n}) = \frac{1}{(2\pi)^N |R_n|} \exp\left(-\frac{1}{2} \underline{n}^T R_n^{-1} \underline{n}\right)$$

where $R_n = \frac{1}{2} E[\underline{n} \underline{n}^T]$

with characteristic function

$$\begin{aligned} M_n(\underline{\omega}) &= E[e^{j \underline{\omega}^T \underline{n}}] \\ &= \exp\left(-j \frac{1}{2} \underline{\omega}^T R_n \underline{\omega}\right) \end{aligned}$$

or with mean $\underline{\mu} = E[\underline{n}]$

$$M_n(\underline{\omega}) = \exp(j \underline{\mu}^T \underline{\omega}) \exp\left(-j \frac{1}{2} \underline{\omega}^T R_n \underline{\omega}\right)$$

What is it
for non-zero
mean?