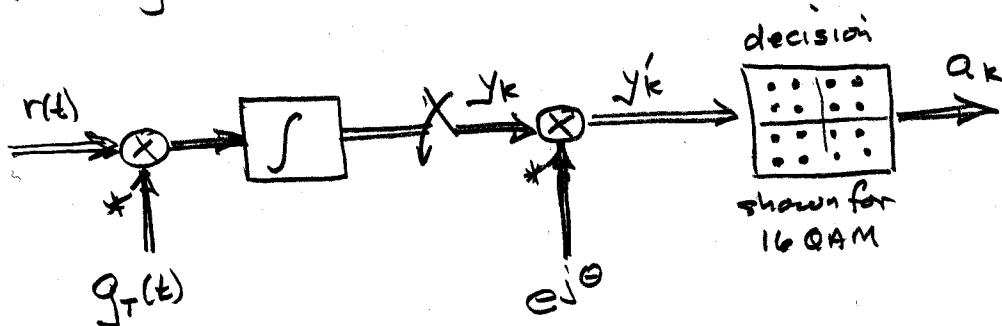


7.4 Coherent and Quasi-Coherent Detection

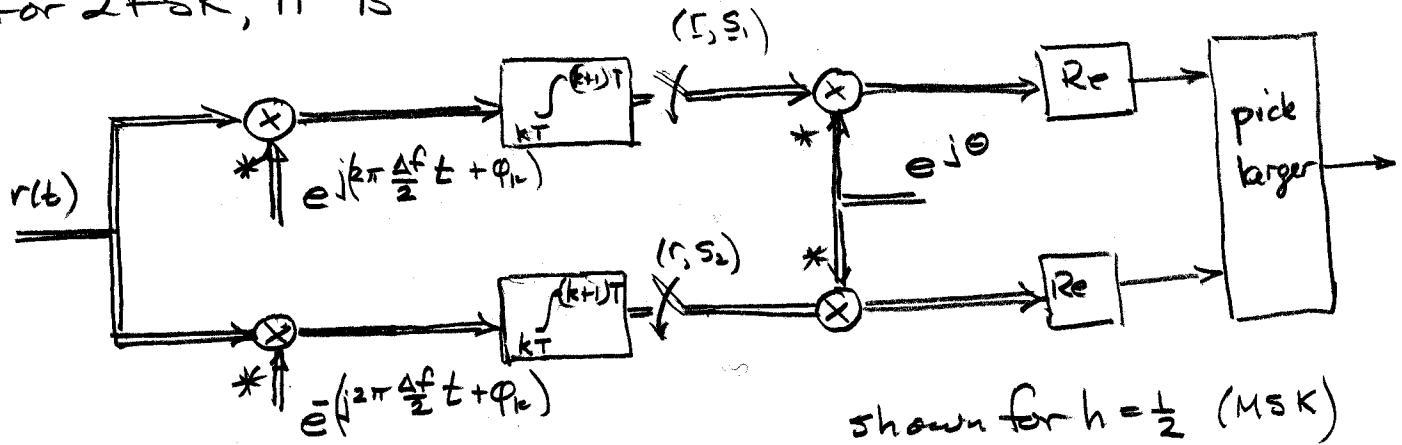
- "coherent detection": you have a perfect phase ref $\hat{\theta} = \theta$ and you use it to de-rotate the received signal
- "quasi-coherent detection": you have a reasonably good phase reference, and you decide to use it as if it were perfect and hope for the best.
- Not too much to say here. Just:
 - structure of coherent receivers
 - performance of coherent receivers
 - justification of the quasi-coherent approach

Coherent Receiver Structures

- For linear modulation it is just what we have been doing



- For 2FSK, it is



- receive $r(t) = e^{j\theta} s_i(t) + n(t)$

lump $e^{j\theta}$ with $s_i(t)$ to make modified signals $s'_i(t) = e^{j\theta} s_i(t)$

- with some orthonormal basis, we have

$$r = s'_i + n$$

- equiprob signals, MAP detector has

$$\hat{i} = \arg \min_i \|r - s'_i\|^2 = \arg \min_i (r - s'_i)^T (r - s'_i)$$

$$= \arg \min_i (|r|^2 - 2 \operatorname{Re}[s'^T r] + |s'_i|^2)$$

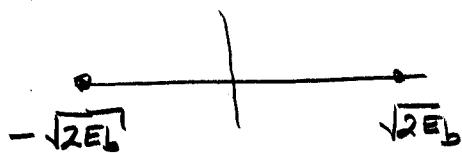
$$= \arg \max_i \operatorname{Re}[r^T s'_i]$$

$$= \arg \max_i \operatorname{Re}\left[e^{-j\theta} \int r(t) s_i^*(t) dt \right]$$

a reminder
that things
are complex

Performance of Coherent Detection

- Coherent detectors eliminate phase offset, so performance is just what we derived in Sections 5 and 6
- Example: BPSK, a binary antipodal set noise variance, each dimension



$$\sigma_v^2 = N_0$$

$$P_b = Q\left(\frac{\sqrt{2E_b}}{\sigma_v}\right) = Q\left(\sqrt{\frac{2E_b}{N_0}}\right) = Q\left(\sqrt{2S_b}\right) \text{ as before.}$$

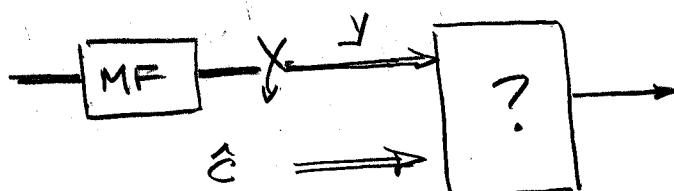
Justification of Quasi-Coherent Structure

- Suppose we follow a feedforward or pilot based approach and create an estimate of the gain $c = e^{j\theta}$

$$\hat{c} = c + w \quad w \text{ is Gaussian noise}$$

$$\text{var } \sigma_w^2 = \frac{1}{2} E[|w|^2]$$

- We have for linear modulation:



- Received sample is $y = c s_i + v$ complex scalar
but $c = \hat{c} - w$, so

$$y = \hat{c} s_i - \underbrace{w s_i + v}_{u, \text{ Gaussian}} = \hat{c} s_i + u$$

$$\sigma_{u,i}^2 = \sigma_w^2 E_i + v$$

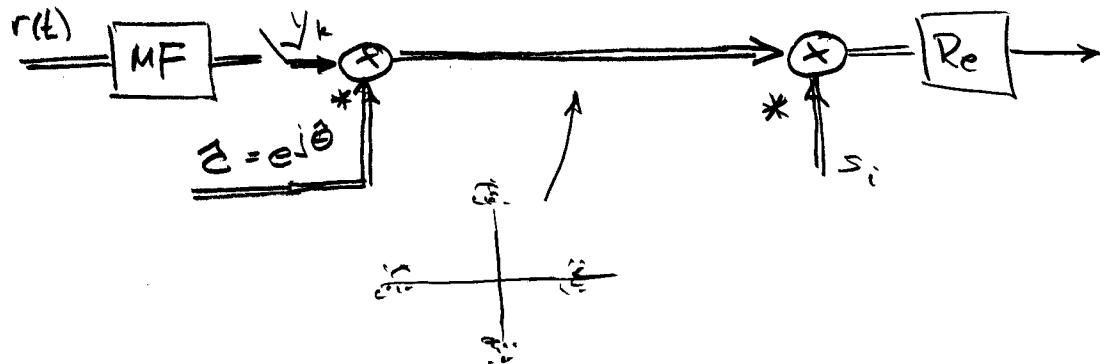
Since we have \hat{c} at hand, we should do

$$\hat{i} = \operatorname{argmax}_i P(y | s_i, \hat{c}) = \frac{1}{2\pi\sigma_{u,i}^2} \exp\left(-\frac{|y - \hat{c}s_i|^2}{2\sigma_{u,i}^2}\right)$$

$$= \operatorname{argmin}_i \left(\frac{|y - \hat{c}s_i|^2}{2\sigma_{u,i}^2} + \ln(\sigma_{u,i}^2) \right) = \operatorname{argmin}_i \left(|y - \hat{c}s_i|^2 + 2\sigma_{u,i}^{-2} \ln(\sigma_{u,i}^2) \right)$$

$$= \operatorname{argmin}_i \left(|y|^2 - 2\Re[y s_i^* \hat{c}^*] + |\hat{c}|^2 s_i^2 + 2\sigma_{u,i}^{-2} \ln(\sigma_{u,i}^2) \right)$$

If equal energy $\hat{i} = \operatorname{argmax}_i \Re[y s_i^* \hat{c}^*]$



- Performance? Difficult integrals of Bessel functions.
Ugh. Closed form in terms of Marcum Q-function.